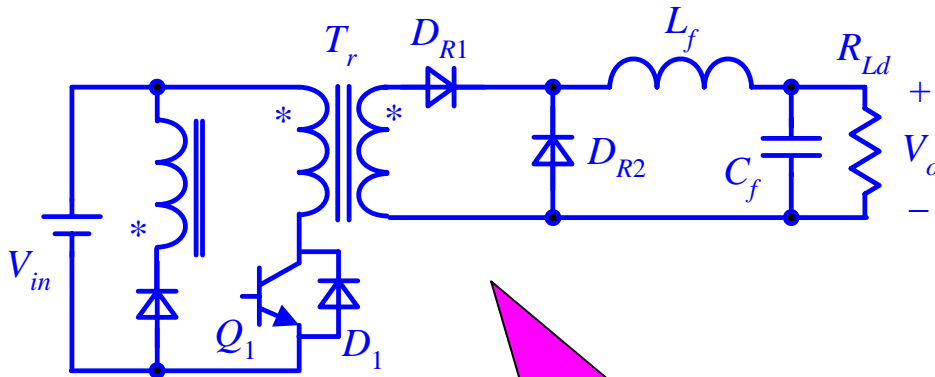


# Buck-Derived Converters

*Presented by*

*Xinbo Ruan*

**Aero-Power Sci-tech Center**  
**Nanjing University of Aeronautics & Astronautics**



It is suited for the low power and low input voltage applications

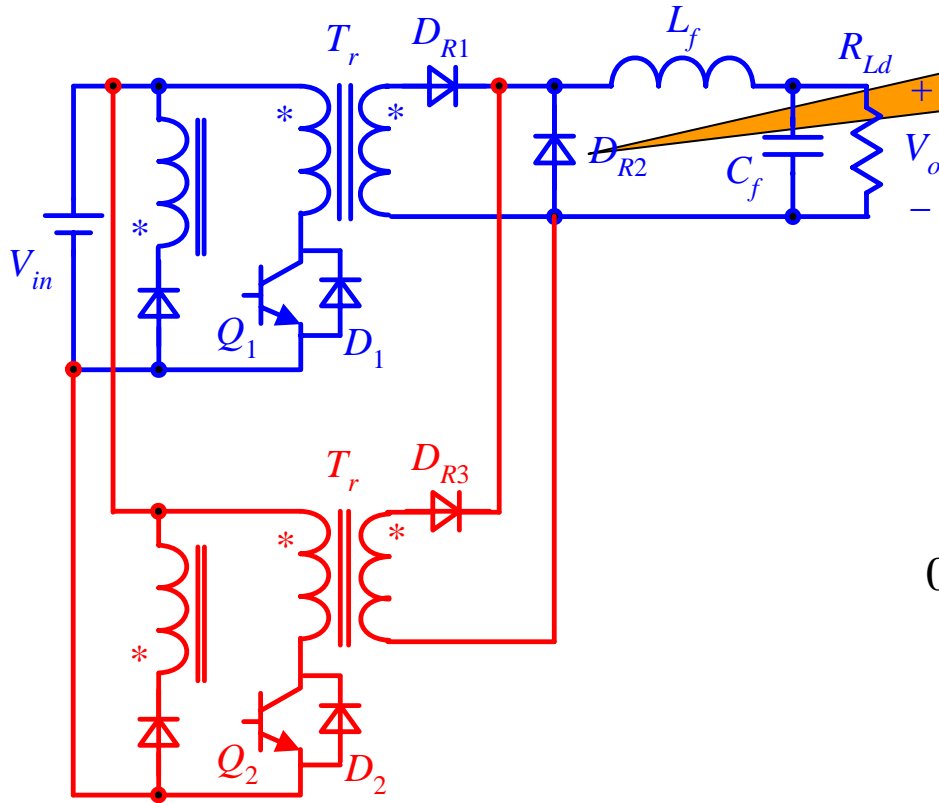
### Advantages:

- ☺ Simple topology;
- ☺ **No direct-short problem.**

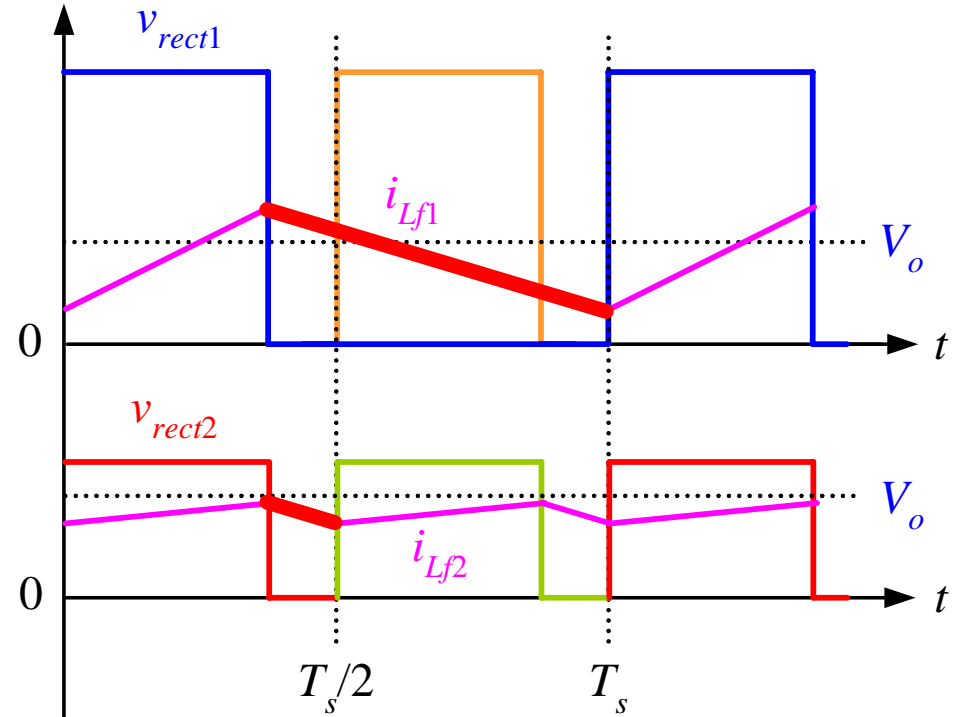
### Disadvantages:

- ☹ The duty cycle is limited to 0.5;
- ☹ The transformer is unidirectional magnetic, so it is very large;
- ☹ The filter inductance is large;
- ☹ The voltage stress of the rectifier diodes is high.

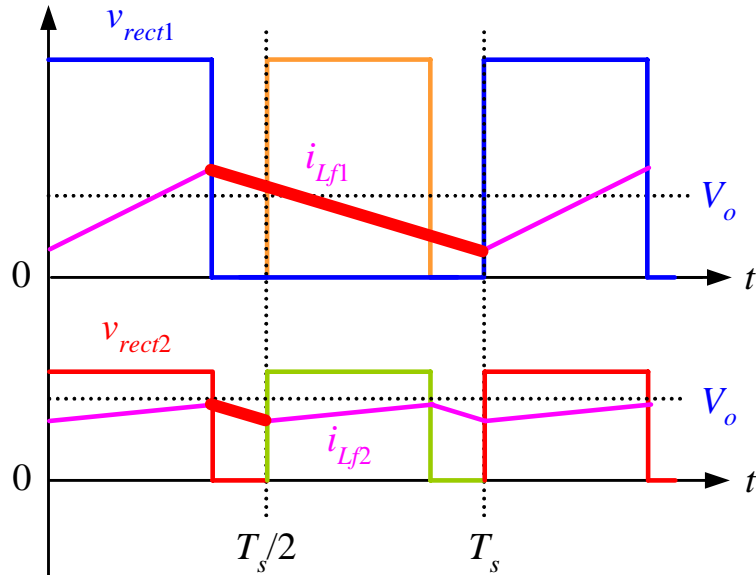
## Double Forward Converter (1)



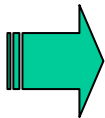
The voltage stress is reduced to half.



# Double Forward Converter (2)




$$V_o = L_f \frac{\Delta i_{L_f}}{T_{off}}$$




$$\Delta i_{L_f} = \frac{V_o}{L_f} \cdot (1 - D) \cdot T_s$$

Given  $D_{max} = 0.45$  for single Forward converter.


$$V_o = 0.45 \cdot \frac{V_{in\ min}}{K}$$



$$D = \frac{KV_o}{V_{in}} = 0.45 \cdot \frac{V_{in\ min}}{V_{in}}$$

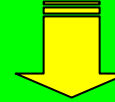


$$\Delta i_{L_f} = \frac{V_o}{L_f} \cdot \left( 1 - 0.45 \cdot \frac{V_{in\ min}}{V_{in}} \right) \cdot T_s$$




$$\Delta i_{L_f}^* = \frac{1 - 0.45 \cdot \frac{V_{in\ min}}{V_{in}}}{1 - 0.45}$$


$$V_o = 0.9 \cdot \frac{V_{in\ min}}{K}$$



$$D = \frac{KV_o}{V_{in}} = 0.9 \cdot \frac{V_{in\ min}}{V_{in}}$$

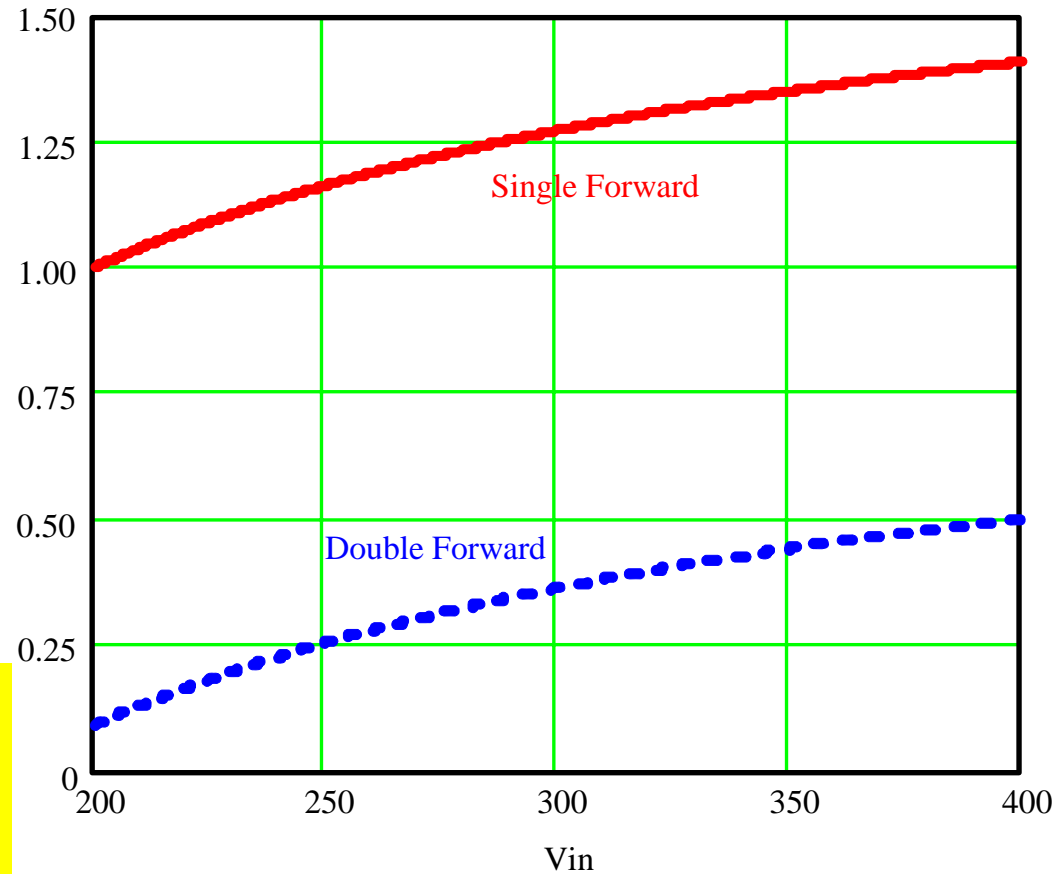
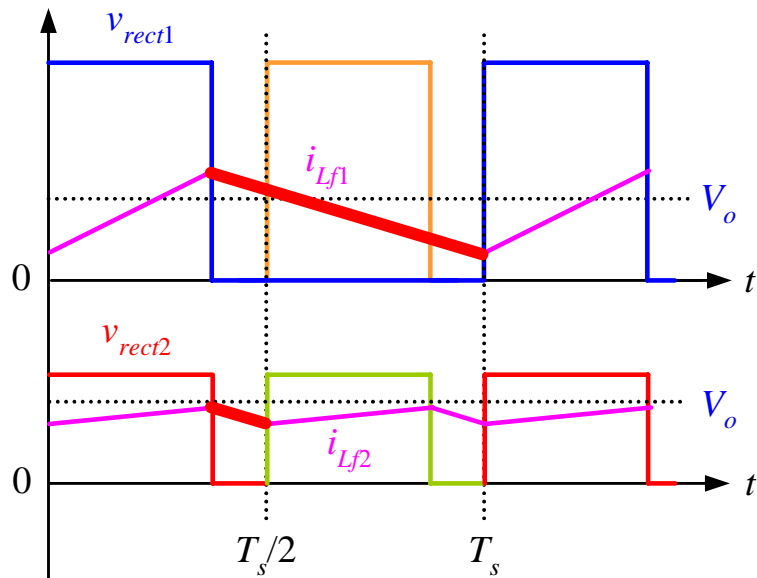


$$\Delta i_{L_f} = \frac{V_o}{L_f} \cdot \left( 1 - 0.9 \cdot \frac{V_{in\ min}}{V_{in}} \right) \cdot \frac{T_s}{2}$$



$$\Delta i_{L_f}^* = \frac{1 - 0.9 \cdot \frac{V_{in\ min}}{V_{in}}}{(1 - 0.45) \times 2}$$

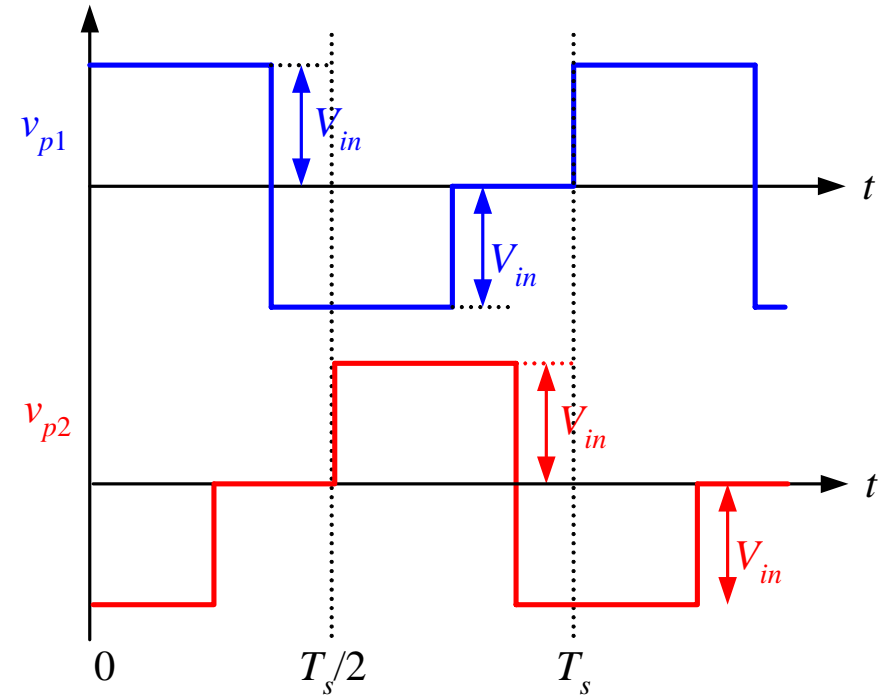
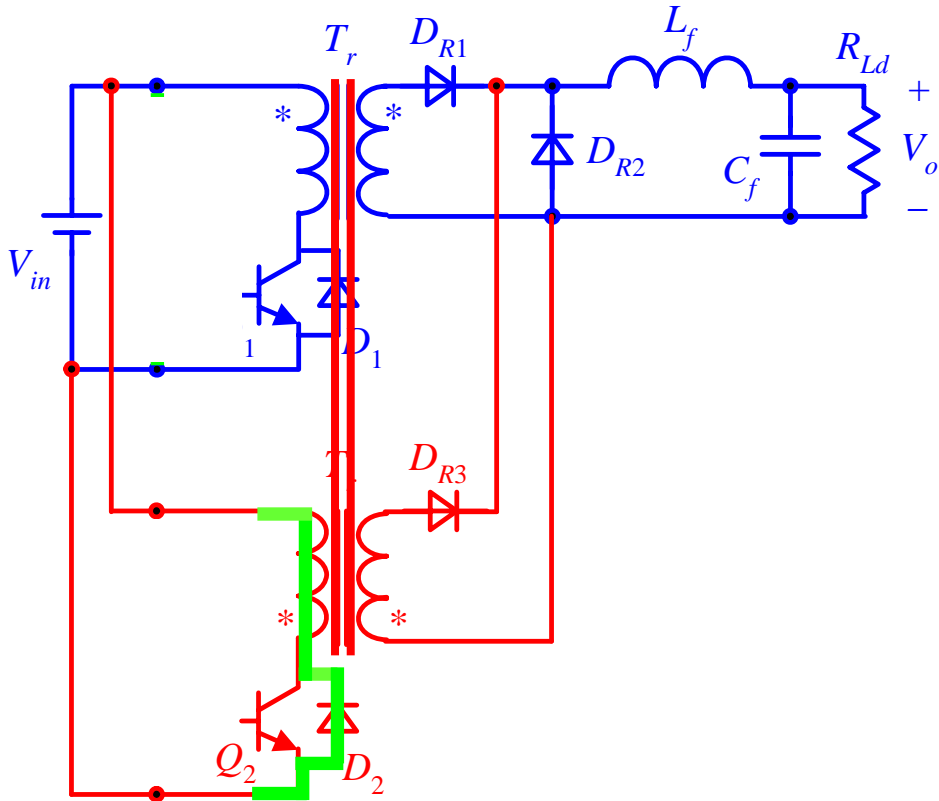
# Double Forward Converter (3)



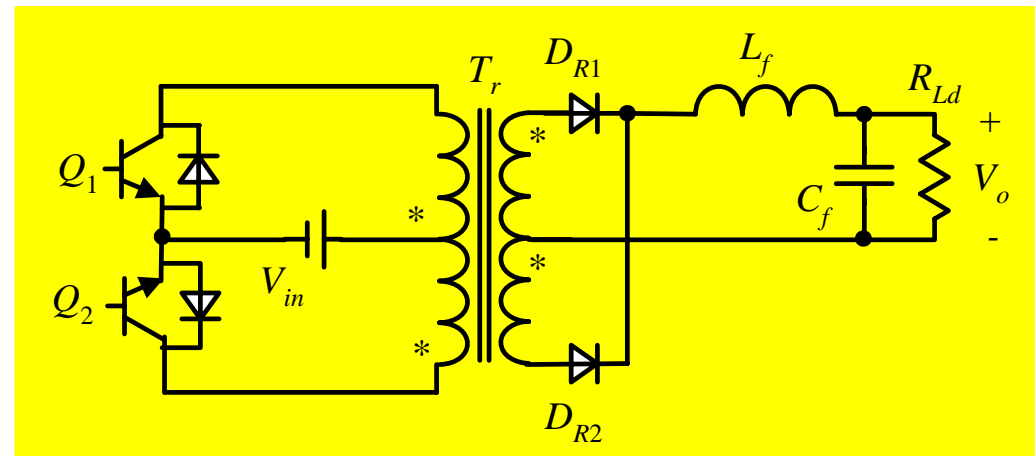
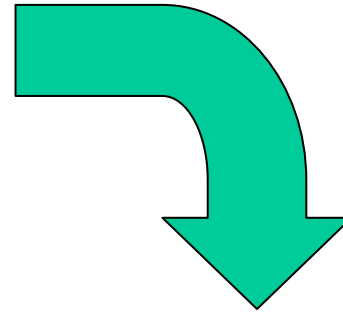
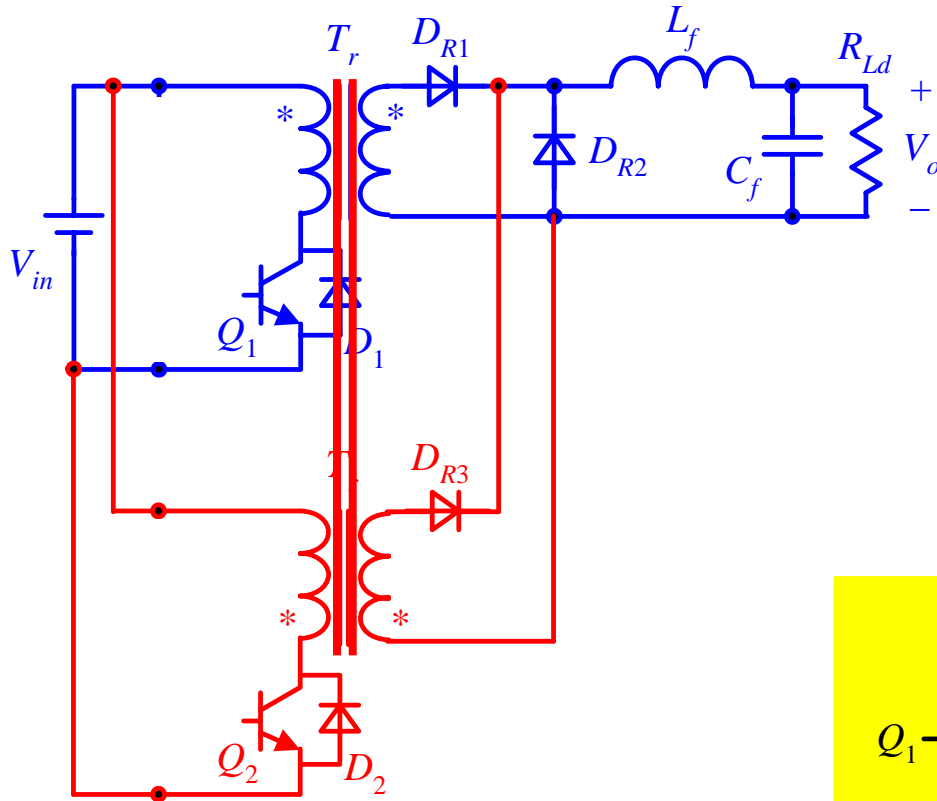
The reasons for the reduction of  $L_f$

- The magnitude of the secondary voltage is reduced to half;
- The ripple frequency is doubled.

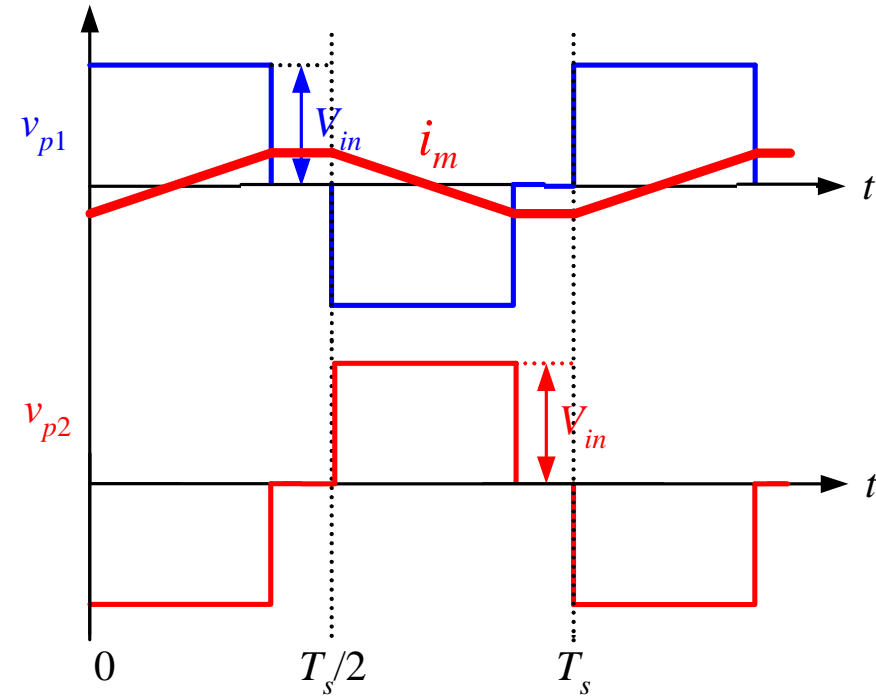
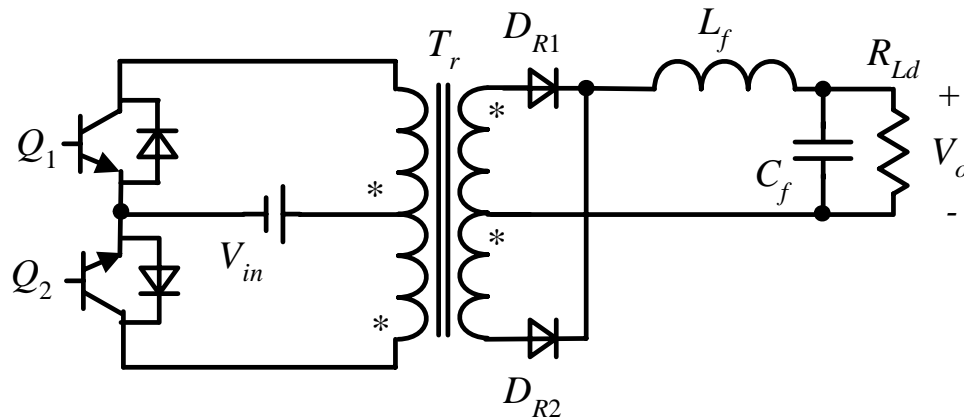
# Derivation of Push-Pull Converter (1)



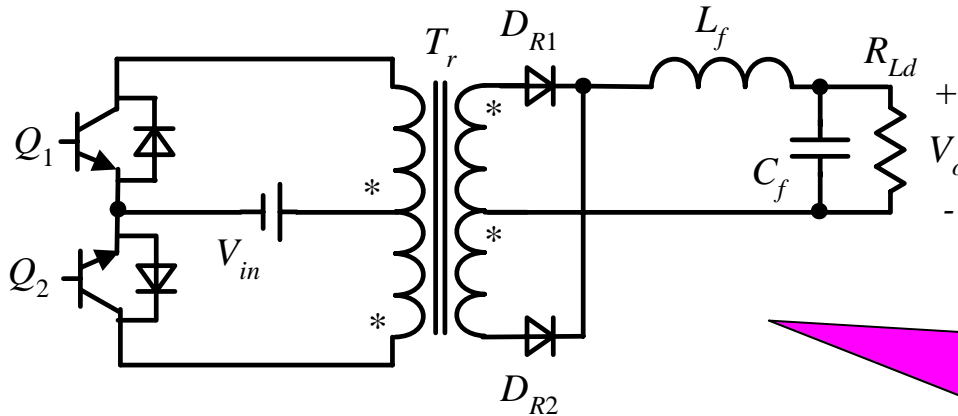
## Derivation of Push-Pull Converter (2)



# Magnetic Reset of Push-Pull Converter







It is suited for the low input voltage applications

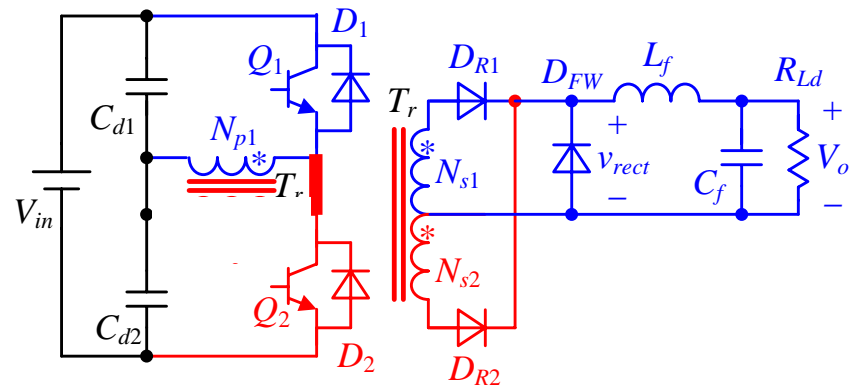
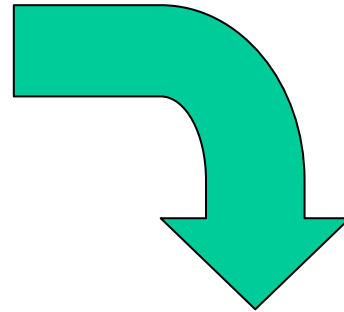
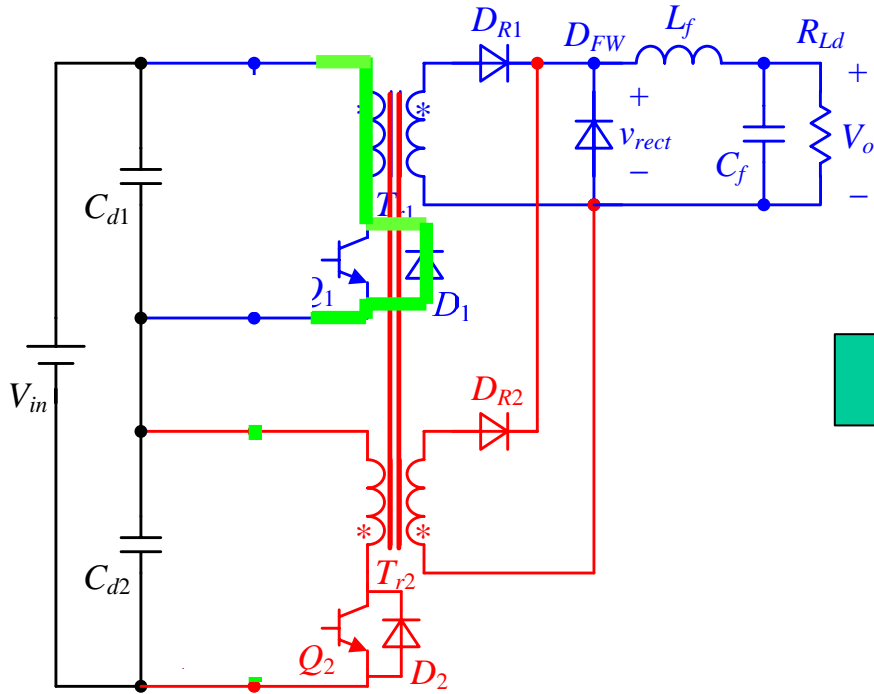
### Advantages:

- ☺ Bi-directional magnetic, leading to small transformer;
- ☺ The duty cycle can be extended to be 1.0, leading to reduced filter inductance and lower voltage stress of the rectifier diodes.

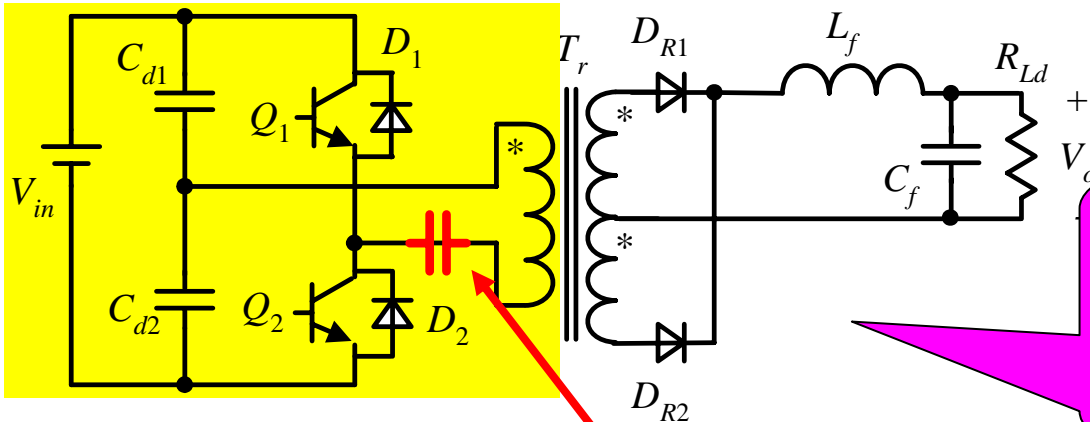
### Disadvantages:

- ☹ The switches sustain twice of the input voltage;
- ☹ The leakage inductance results in large voltage spike across the switches;
- ☹ There is a risk of saturation of the transformer due to the dc component.

# Derivation of Half-Bridge Converter



## Half-Bridge Converter



medium

It is suited for the low input voltage applications

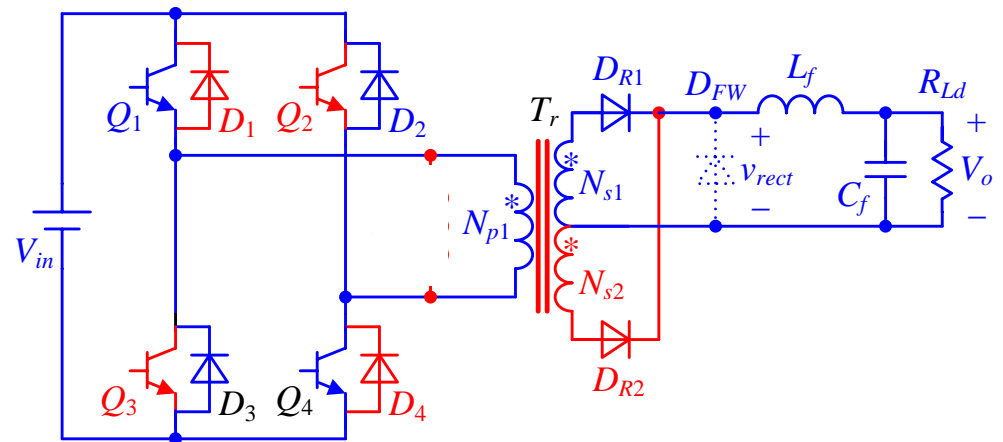
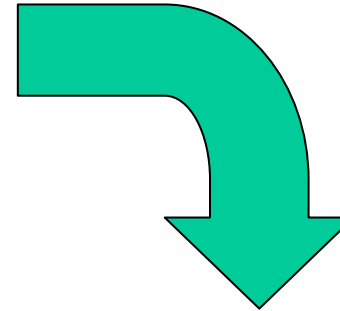
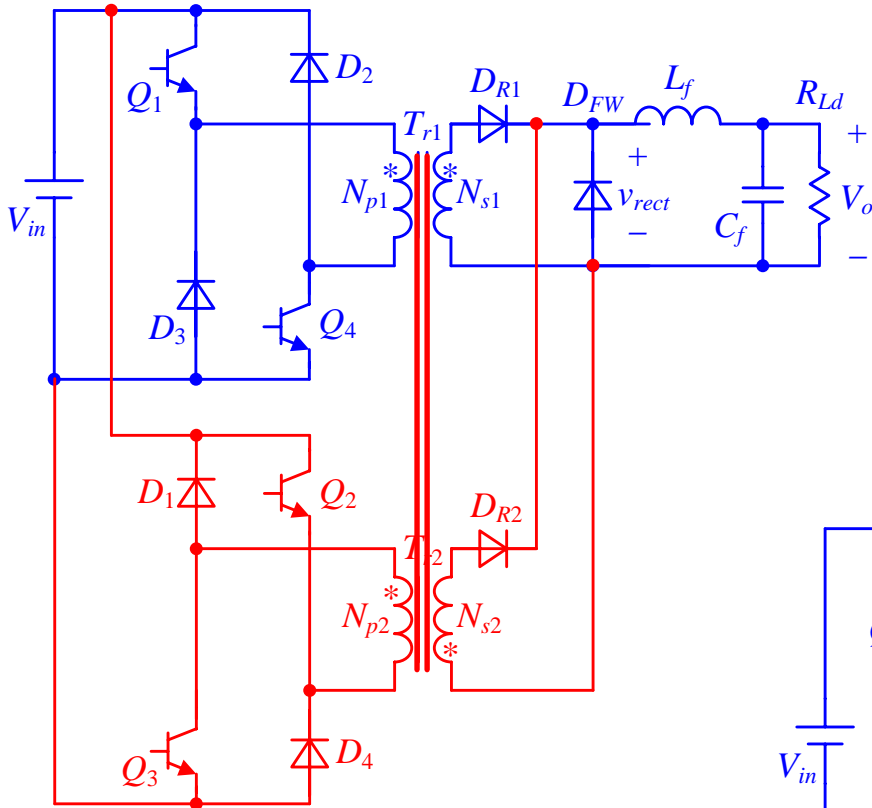
### Advantages:

- ☺ Bi-directional magnetic, leading to small transformer;
- ☺ The duty cycle can be extended to be 1.0, leading to reduced filter inductance and lower voltage stress of the rectifier diodes.

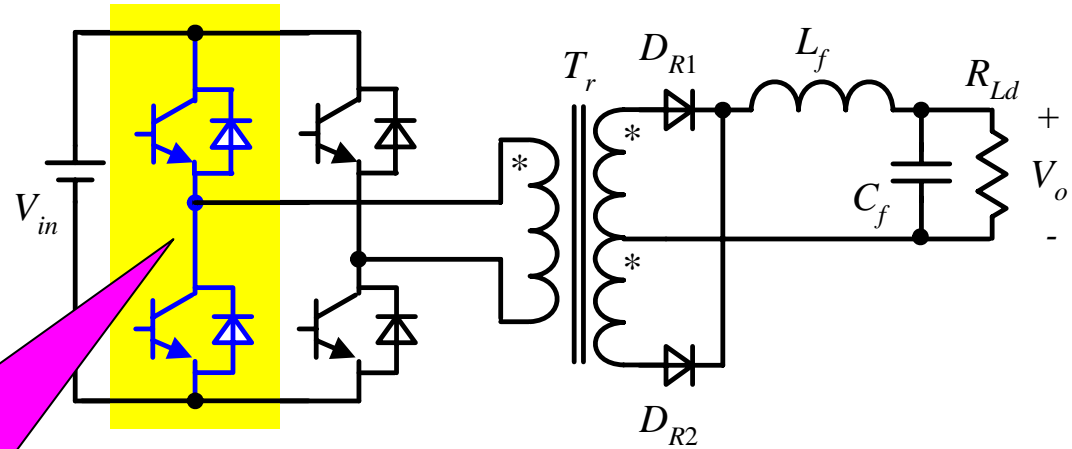
### Disadvantages:

- ☹ The switches sustain ~~twice of~~ the input voltage;
- ☹ ~~The leakage inductance results in large voltage spike across the switches;~~
- ☹ There is a risk of saturation of the transformer due to the dc component.

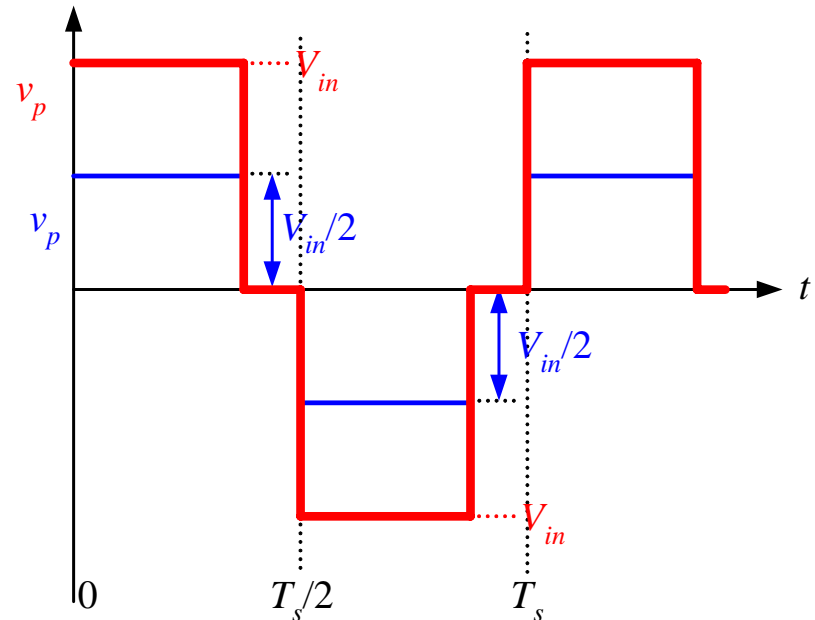
# Derivation of Full-Bridge Converter



## Full-Bridge Converter



The current stress of the switches reduces to be half compared with HB

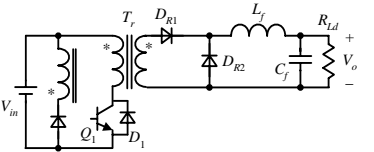
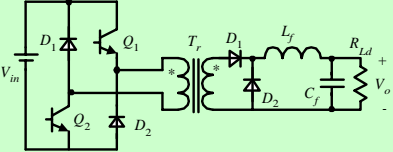
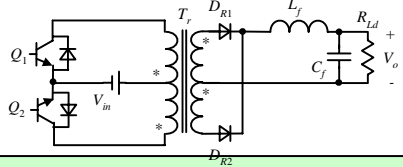
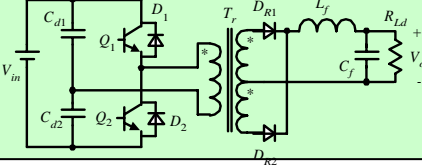
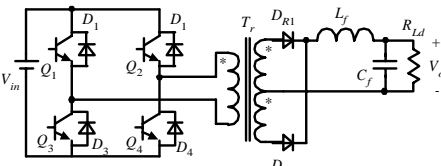


## Design a DC/DC converter

- Input Voltage:  $V_{in}$
- Output Voltage:  $V_o$
- Output Current:  $I_o$

Given the duty cycle of all the switches be 0.5

# Comparison of the Converters

Converter	Topology	Winding Ratio	Current Stress	Voltage Stress	No of Switches	Total Power of Switches
Single-Switch Forward		$K_0$	$I_o/K_0$	$2V_{in}$	1	$2V_{in}I_o/K_0$
Dual-Switch Forward		$K_0$	$I_o/K_0$	$V_{in}$	2	$2V_{in}I_o/K_0$
Push-Pull		$2K_0$	$I_o/(2K_0)$	$2V_{in}$	2	$2V_{in}I_o/K_0$
Half-Bridge		$K_0$	$I_o/K_0$	$V_{in}$	2	$2V_{in}I_o/K_0$
Full-Bridge		$2K_0$	$I_o/(2K_0)$	$V_{in}$	4	$2V_{in}I_o/K_0$

*Thanks for your attention !*

*Questions? / Answer!*